

Laplace Transforms

- 2-sided or bilateral Laplace transform

$$x(t) \xleftrightarrow{\mathcal{L}} X(s) = \int_{-\infty}^{\infty} x(t)e^{-st} dt$$

ROC : $s_a < \text{Re}\{s\} < s_b \Rightarrow$ region of convergence $\Rightarrow s$ -values for which the integral converges

ROC \Rightarrow vertical strip in the complex plane

- $e^{s_0 t}$ are not Laplace Transformable
- If $x(t)$ is right-sided ($x(t < t_0) = 0$),
then $s_b = \infty \Rightarrow \text{ROC} \Rightarrow s_a < \text{Re}\{s\} < \infty$
- If $x(t)$ is left-sided ($x(t > t_0) = 0$),
then $s_a = -\infty \Rightarrow \text{ROC} \Rightarrow -\infty < \text{Re}\{s\} < s_b$
- If $x(t)$ is time-limited,
then ROC $\Rightarrow -\infty < \text{Re}\{s\} < \infty$

| $x(t)$ | $X(s)$ | ROC $\Rightarrow \text{Re}\{s\} \hat{I}$ |
|---------------------------|---------------------------|---|
| $c_1 x_1(t) + c_2 x_2(t)$ | $c_1 X_1(s) + c_2 X_2(s)$ | (usually) $(\text{ROC})_1 \cap (\text{ROC})_2$ |

Linearity

Need $(\text{ROC})_1 \cap (\text{ROC})_2 \neq \emptyset$

| | | |
|--------------|-------------------|----------|
| $x(t - t_0)$ | $e^{-s t_0} X(s)$ | same ROC |
|--------------|-------------------|----------|

Time-shift rule

| | | |
|---------------------|---------|--|
| $\frac{d}{dt} x(t)$ | $sX(s)$ | (usually) same or at least, contain $(\text{ROC})_X$ |
|---------------------|---------|--|

Time-derivative rule

| | | |
|-------------------|-----------------------|--------------------------------------|
| $x_1(t) * x_2(t)$ | $X_1(s) \cdot X_2(s)$ | $(\text{ROC})_1 \cap (\text{ROC})_2$ |
|-------------------|-----------------------|--------------------------------------|

Convolution rule

Need $(\text{ROC})_1 \cap (\text{ROC})_2 \neq \emptyset$

| | | |
|--------------------|---------------------|-------------------------------|
| $e^{s_0 t} u(t)$ | $\frac{1}{s - s_0}$ | $(\text{Re}\{s_0\}, \infty)$ |
| $-e^{s_0 t} u(-t)$ | $\frac{1}{s - s_0}$ | $(-\infty, \text{Re}\{s_0\})$ |

Proof
$$\int_{-\infty}^{\infty} e^{s_0 t} u(t) e^{-st} dt = \int_0^{\infty} e^{s_0 t} e^{-st} dt = \int_0^{\infty} e^{-(s-s_0)t} dt = \frac{1}{s - s_0}$$

Need $\text{Re}\{s - s_0\} > 0$ or $\text{Re}\{s\} > \text{Re}\{s_0\}$

Proof
$$\int_{-\infty}^{\infty} -e^{s_0 t} u(-t) e^{-st} dt = - \int_{-\infty}^0 e^{s_0 t} e^{-st} dt = - \int_{-\infty}^0 e^{(s_0 - s)t} dt = - \frac{1}{s_0 - s} = \frac{1}{s - s_0}$$

Need $\text{Re}\{s_0 - s\} > 0$ or $\text{Re}\{s\} < \text{Re}\{s_0\}$

| | | |
|--|----------------------------------|---|
| $t e^{s_0 t} u(t)$ | $\frac{1}{(s - s_0)^2}$ | $(\text{Re}\{s_0\}, \infty)$ |
| $-t e^{s_0 t} u(-t)$ | | $(-\infty, \text{Re}\{s_0\})$ |
| $\frac{t^k}{k!} e^{s_0 t} u(t)$ | $\frac{1}{(s - s_0)^{k+1}}$ | $(\text{Re}\{s_0\}, \infty)$ |
| $-\frac{t^k}{k!} e^{s_0 t} u(-t)$ | | $(-\infty, \text{Re}\{s_0\})$ |
| $u(t)$ | $\frac{1}{s}$ | $(0, \infty)$ |
| $-u(-t)$ | | $(-\infty, 0)$ |
| $\mathbf{d}(t)$ | 1 | $(-\infty, \infty)$ |
| $\mathbf{d}(t-T)$ | e^{-sT} | $(-\infty, \infty)$ |
| $P_T(t)$ | $\frac{e^{sT} - e^{-sT}}{s}$ | $(-\infty, \infty)$ |
| $u(t) \cos(\mathbf{w}_0 t)$ | $\frac{s}{s^2 - \mathbf{w}_0^2}$ | $(0, \infty)$ |
| $e^{-s_0 t } = e^{-s_0 t} u(t) + e^{s_0 t} u(-t)$ | $\frac{-2s_0}{s^2 - s_0^2}$ | $(-\text{Re}\{s_0\}, \text{Re}\{s_0\})$ |

- \mathcal{L} -transform and \mathfrak{F} -transform
 - If imaginary axis lies inside $(\text{ROC})_X$ ($\sigma_a < 0 < \sigma_b$), then $\hat{X}(\mathbf{w})$ exists and $= X(s = j\mathbf{w})$
 - If the imaginary axis is bounded away from $(\text{ROC})_X$, then, $x(t)$ has no \mathfrak{F} -transform
 - If the imaginary axis is an "edge" of $(\text{ROC})_X$, then, $x(t)$ may or may not have a \mathfrak{F} -transform
 - won't in general have $\hat{X}(\mathbf{w}) = X(s = j\mathbf{w})$
 - $\hat{X}(\mathbf{w})$ will usually contain \mathbf{d}

Recovering $x(t)$ from $X(s)$, $(\text{ROC})_X$

- Pick $\hat{\mathbf{s}}$ so that line $s = \hat{\mathbf{s}} + j\omega$ is inside $(\text{ROC})_X$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\hat{s} + j\omega) e^{(\hat{s} + j\omega)t} d\omega$$

- Given $X(s) = \frac{p(s)}{q(s)}$

roots of $p(s) \Rightarrow$ **zeros** of $X(s)$

roots of $q(s) \Rightarrow$ **poles** of $X(s)$

- No poles of $X(s)$ lies in $(\text{ROC})_X$
- Finite edge(s) of $(\text{ROC})_X$ pass through one or more poles

- Partial fraction

- $X(s) = \frac{r_1(s)}{r_2(s)} \frac{1}{s - s_0}$; $r_1(s_0), r_2(s_0) \neq 0$

$$X(s) = \frac{K}{s - s_0} + r_3(s) \text{ where } K = \lim_{s \rightarrow s_0} ((s - s_0) X(s))$$

Proof $\frac{r_1(s)}{r_2(s)} \frac{1}{s - s_0} = \frac{K}{s - s_0} + r_3(s) \Rightarrow \frac{r_1(s)}{r_2(s)} = K + (s - s_0)r_3(s)$

$$\frac{r_1(s_0)}{r_2(s_0)} = K + 0 = K \Rightarrow K = \frac{r_1(s_0)}{r_2(s_0)} = \lim_{s \rightarrow s_0} ((s - s_0) X(s))$$

- $X(s) = \frac{r_1(s)}{r_2(s)} \frac{1}{(s - s_0)^j}$; $r_1(s_0), r_2(s_0) \neq 0$

$$X(s) = \sum_{m=1}^j \frac{K_m}{(s - s_0)^m} + r_3(s) \text{ where}$$

$$K_j = \lim_{s \rightarrow s_0} ((s - s_0)^j X(s))$$

$$K_m = \frac{1}{(j - m)!} \lim_{s \rightarrow s_0} \left(\frac{d^{(j-m)}}{ds^{(j-m)}} ((s - s_0)^j X(s)) \right)$$

Proof $\frac{r_1(s)}{r_2(s)} \frac{1}{(s - s_0)^j} = \sum_{m=1}^j \frac{K_m}{(s - s_0)^m} + r_3(s)$

$$\Rightarrow \frac{r_1(s)}{r_2(s)} = \sum_{m=1}^j K_m (s - s_0)^{j-m} + (s - s_0)^j r_3(s)$$

$$= \sum_{m=1}^{j-1} K_m (s - s_0)^{j-m} + K_j (s - s_0)^{0^1} + (s - s_0)^j r_3(s)$$

$$\frac{r_1(s_0)}{r_2(s_0)} = K_j$$

$$\frac{d^n r_1(s)}{ds^n r_2(s)} = \sum_{m=1}^j K_m \frac{d^n}{ds^n} (s-s_0)^{j-m} + \frac{d^n}{ds^n} (s-s_0)^j r_3(s) \quad ; n < j-m$$

$$\text{Note that } \frac{d^n}{ds^n} (s-s_0)^k = \begin{cases} 0 & \text{when } n > k \\ k! & \text{when } n = k \\ \frac{k!}{(k-n)!} (s-s_0)^{k-n} & \text{when } n < k \end{cases} .$$

$$\text{So, } \left(\frac{d^n}{ds^n} (s-s_0)^k \right)_{s=s_0} = \begin{cases} k! & \text{when } n = k \\ 0 & \text{otherwise} \end{cases} .$$

$$\text{Similarly, } \left(\frac{d^n}{ds^n} (s-s_0)^{j-m} \right)_{s=s_0} = \begin{cases} (j-m)! & \text{when } n = j-m \\ 0 & \text{otherwise} \end{cases}$$

$$\text{Thus, } \left(\frac{d^n r_1(s)}{ds^n r_2(s)} \right)_{s=s_0} = (\dots + 0 + K_{j-n} (n!) + 0 + \dots) + 0 = K_{j-n} (n!)$$

$$\text{And } K_{j-n} = \frac{1}{n!} \left(\frac{d^n r_1(s)}{ds^n r_2(s)} \right)_{s=s_0} \Rightarrow K_m = \frac{1}{(j-m)!} \left(\frac{d^{(j-m)} r_1(s)}{ds^{(j-m)} r_2(s)} \right)_{s=s_0}$$

- $X(s) = \frac{p(s)}{\prod_i (s-s_i)} = \sum_i \frac{K_i}{s-s_i}$ where $K_i = \lim_{s \rightarrow s_i} ((s-s_i)X(s))$

- $\frac{1}{s^2+a^2} = \frac{1}{2ia} \left(\frac{1}{s-ia} - \frac{1}{s+ia} \right)$

$$\frac{1}{s^2-a^2} = \frac{1}{2a} \left(\frac{1}{s-a} - \frac{1}{s+a} \right)$$

$$\frac{s}{s^2-a^2} = \frac{1}{2} \left(\frac{1}{s-a} + \frac{1}{s+a} \right)$$

$$\frac{s}{s^2+a^2} = \frac{1}{2} \left(\frac{1}{s-ia} + \frac{1}{s+ia} \right)$$

- $\frac{s}{(s-a)^2} = \frac{1}{s-a} + \frac{a}{(s-a)^2}$

$$\frac{1}{(s-a)(s-b)} = \frac{1}{s-a} - \frac{1}{s-b}$$

$$\frac{s}{(s-a)(s-b)} = \frac{a}{s-a} - \frac{b}{s-b}$$

- If $X(s)$ is a strictly proper rational function of s
 - Graph $(\text{ROC})_X$ and mark the poles of $X(s)$
 - Expand $X(s) = \frac{p(s)}{q(s)}$ (degree $p < q$) in partial fractions
 - use prototypes
 - Leftward poles \rightarrow the $u(t)$ -part of $x(t)$
 - Rightward poles \rightarrow the $u(-t)$ -part of $x(t)$

- If know $z(t) \xleftrightarrow{\mathcal{L}} \frac{P(s)}{q(s)}$, then $z(t-T) \xleftrightarrow{\mathcal{L}} e^{-sT} \frac{P(s)}{q(s)}$

Proof From $x(t-T) \xleftrightarrow{\mathcal{L}} e^{-sT} X(s)$

- If degree $p = \text{degree } q$,
write $Y(s) = K_0 + \frac{p_1(s)}{q(s)}$. Use $K_0 \mathbf{d}(t) \xleftrightarrow{\mathcal{L}} K_0$
- For $x(t)$ to be **causal** and have $X(s)$ as "the formula part" of its \mathcal{L} -transform, need all poles of $X(s)$ to be to the left of $(\text{ROC})_X \Rightarrow$ have only $u(t)$ -terms, no $u(-t)$ -term

- $x(t) = \sum_{\ell} k_{\ell} e^{s_{\ell} t} u(t) \xleftrightarrow{\mathcal{L}} X(s) = \sum_{\ell} \frac{k_{\ell}}{s - s_{\ell}} ; \text{Re}\{s\} > \max_{\ell} \text{Re}\{s_{\ell}\}$

$$(\text{ROC})_X = \bigcap_{\ell} (\text{Re}\{s_{\ell}\}, \infty) = (\max_{\ell} \text{Re}\{s_{\ell}\}, \infty)$$

- For 1) rational $H(s)$ 2) causal system,
 $(\text{ROC})_H =$ the part of complex plane to the right of all poles of $H(s)$

One-sided Laplace transform

- $x(t) \xleftrightarrow{\mathcal{L}} X_I(s) = \int_0^{\infty} x(t) u(t) e^{-st} dt$

- $\sigma_b \rightarrow +\infty$

- $Dx(t) \xleftrightarrow{\mathcal{L}} X_I(s) - x(0^-)$

$$D^2 x(t) \xleftrightarrow{\mathcal{L}} s(X_I(s) - x(0^-)) - Dx(0^-)$$

$$\xleftrightarrow{\mathcal{L}} s^2 X_I(s) - s \cdot x(0^-) - Dx(0^-)$$

$$D^3 x(t) \xrightarrow{\mathcal{L}} s(s^2 X_I(s) - s \cdot x(0^-) - Dx(0^-)) - D^2 x(0^-)$$

- To convert $Y_I(s)$ back to $y(t)$

$(ROC)_{Y_I} \Rightarrow$ region to the right of all poles of $Y_I(s)$

Follow the same recipe as the inverse Laplace transform \Rightarrow get $y(t)u(t) \Rightarrow$ cancel $u(t)$
 \Rightarrow get $y(t)$ for $t \geq 0$